

# Digital Communication Systems

## ECS 452

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### 5.2 Binary Convolutional Codes



#### Office Hours:

BKD, 6th floor of Sirindhralai building

**Monday**                    10:00-10:40

**Tuesday**                    12:00-12:40

**Thursday**                    14:20-15:30

# Binary Convolutional Codes

- Introduced by Elias in 1955
  - There, it is referred to as convolutional parity-check symbols codes.
  - Peter Elias received
    - Claude E. Shannon Award in 1977
    - IEEE Richard W. Hamming Medal in 2002
      - for "fundamental and pioneering contributions to information theory and its applications"
- The encoder **has memory**.
  - In other words, the encoder is a **sequential circuit** or a **finite-state machine**.
    - Easily implemented by shift register(s).
    - The **state** of the encoder is defined as **the contents of its memory**.

# Binary Convolutional Codes

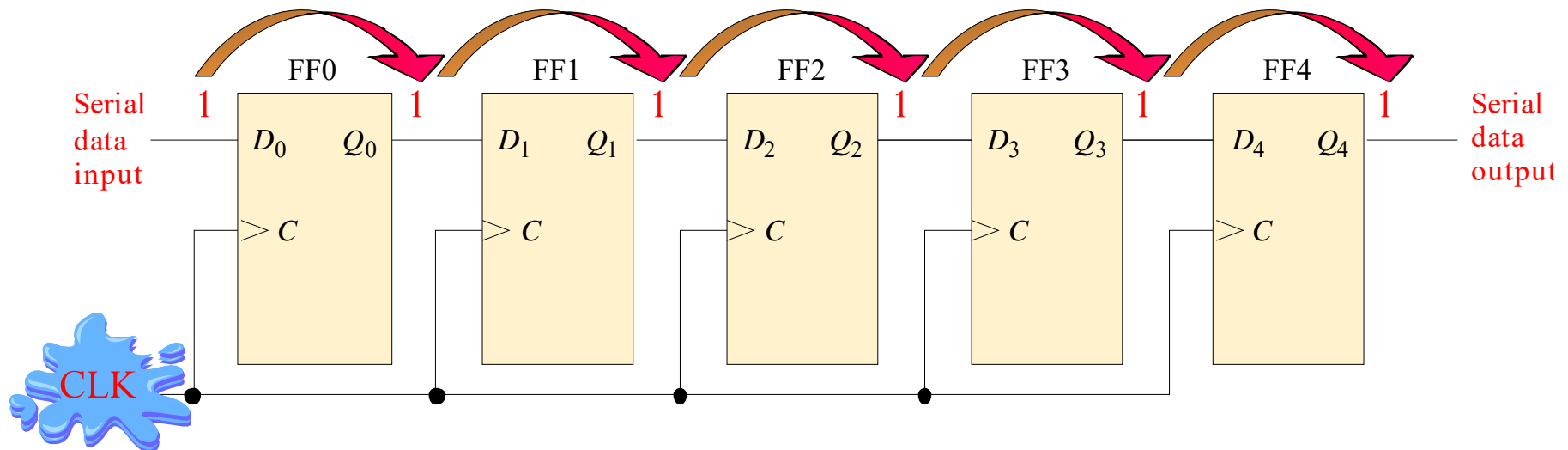
- The encoding is done on a **continuous** running basis rather than by blocks of  $k$  data digits.
  - So, we use the terms **bit streams** or **sequences** for the input and output of the encoder.
  - In theory, these sequences have infinite duration.
  - In practice, the state of the convolutional code is periodically forced to a known state and therefore code sequences are produced in a block-wise manner.

# Binary Convolutional Codes

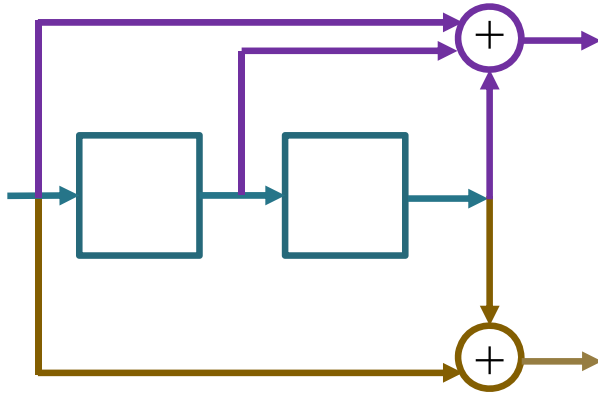
- In general, a **rate- $\frac{k}{n}$  convolutional encoder** has
  - $k$  shift registers, one per input information bit, and
  - $n$  output coded bits that are given by linear combinations (over the binary field, of the contents of the registers and the input information bits.
- $k$  and  $n$  are usually small.
- For simplicity of exposition, and for practical purposes, only **rate- $\frac{1}{n}$**  binary convolutional codes are considered here.
  - $k = 1$ .
  - These are the most widely used binary codes.

# (Serial-in/Serial-out) Shift Register

- Accept data serially: one bit at a time on a single line.
- Each clock pulse will move an input bit to the next FF.  
For example, a 1 is shown as it moves across.
- Example: five-bit serial-in serial-out register.



# Example 1: $n = 2, k = 1$

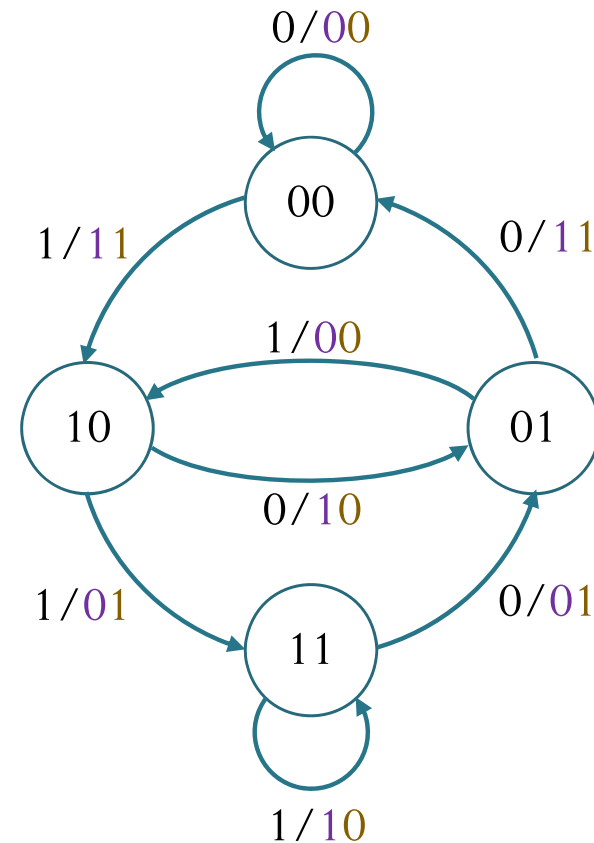
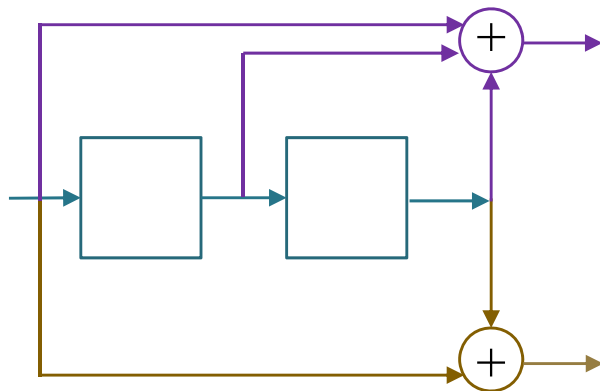


# Graphical Representations

- Three different but related graphical representations have been devised for the study of convolutional encoding:
  1. the state diagram
  2. the code tree
  3. the trellis diagram

# Ex. 1: State (Transition) Diagram

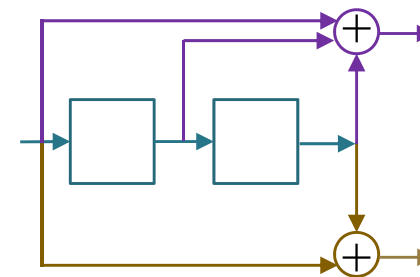
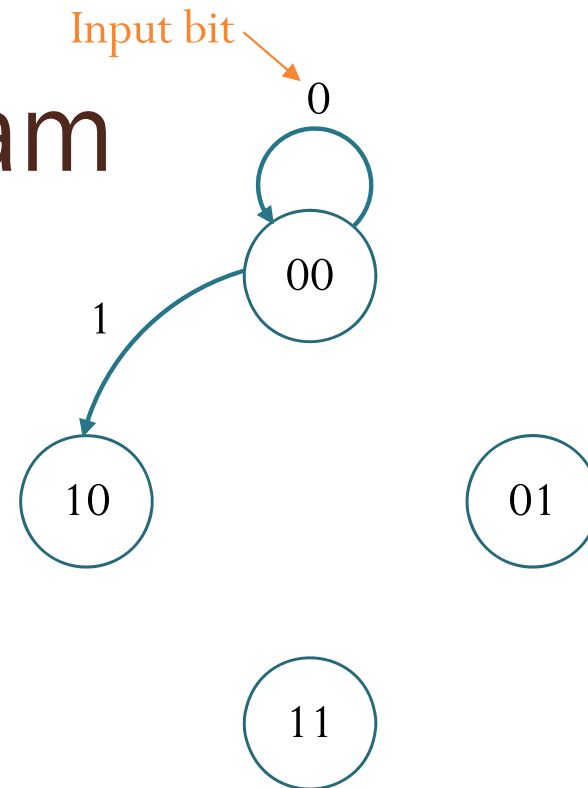
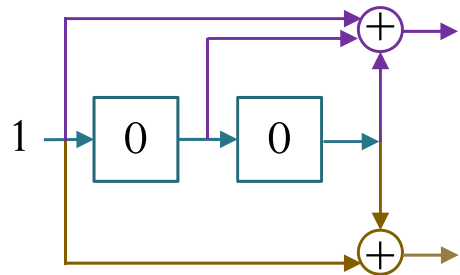
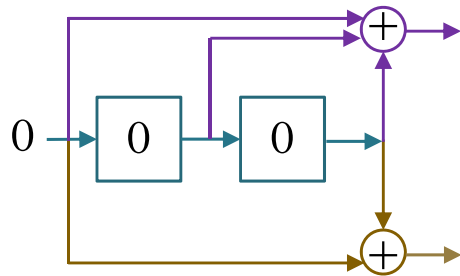
- The encoder behavior can be seen from the perspective of a finite state machine with its state (transition) diagram.



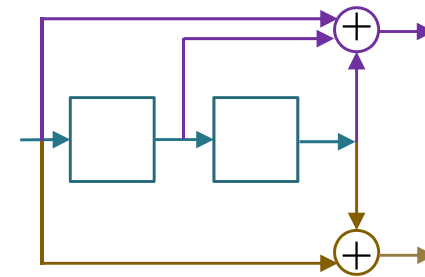
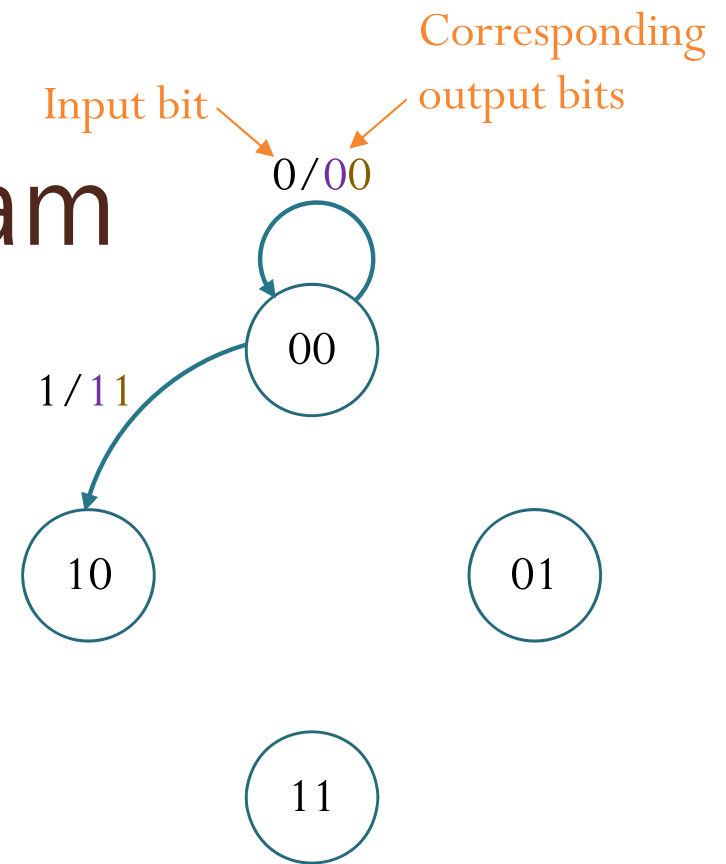
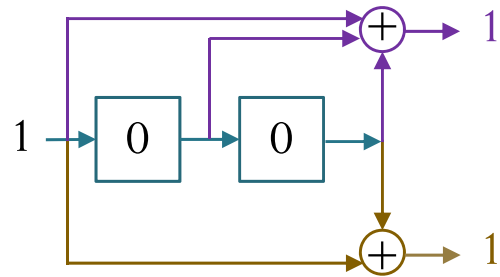
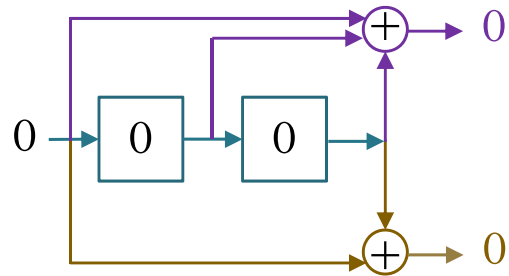
A four-state directed graph that uniquely represents the input-output relation of the encoder.



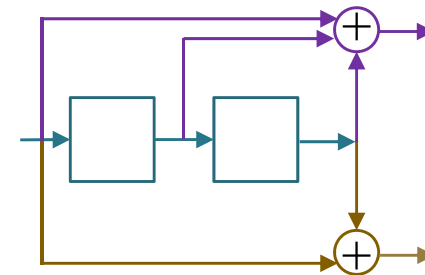
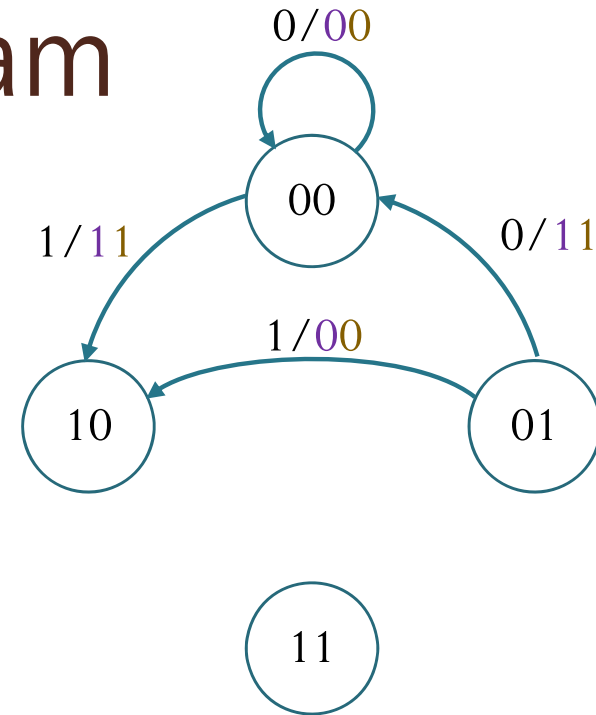
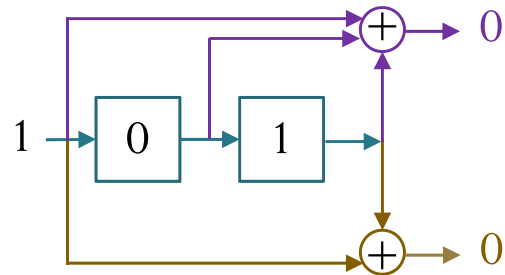
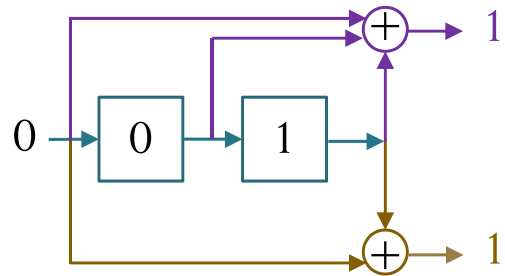
# Drawing State Diagram



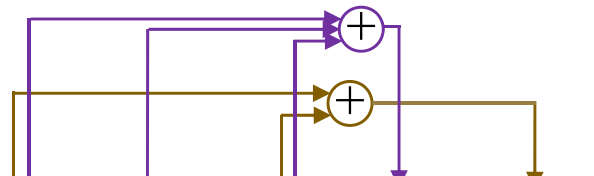
# Drawing State Diagram



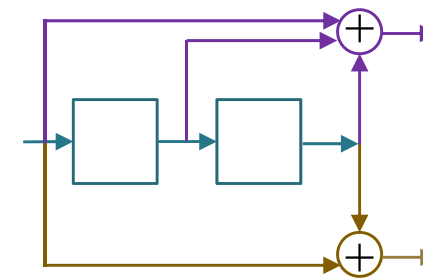
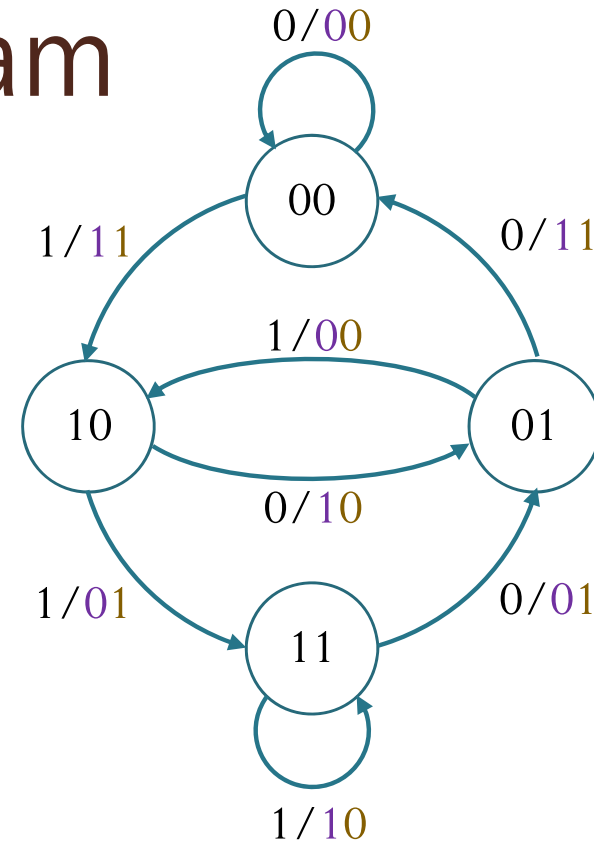
# Drawing State Diagram



# Drawing State Diagram

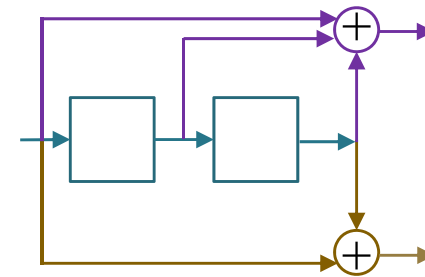
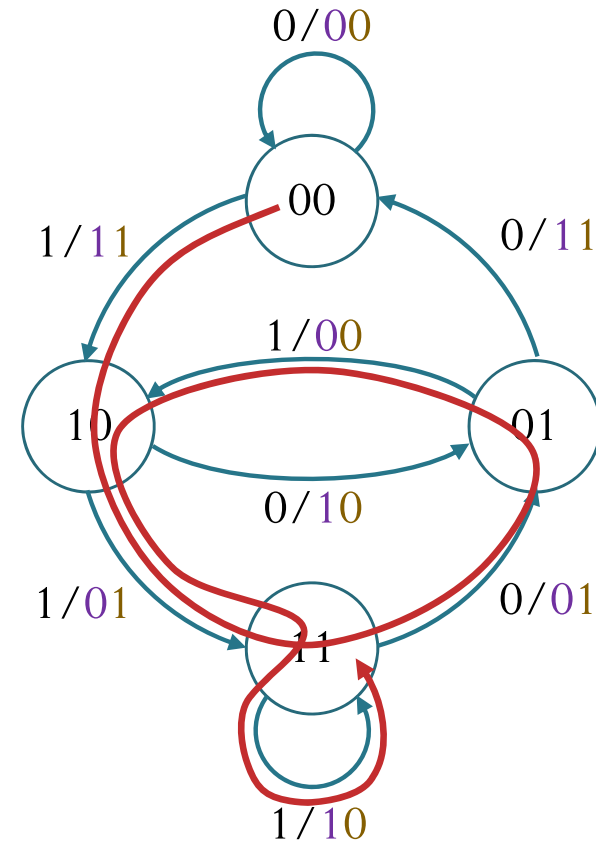


$b$	$s_1$	$s_2$	$x^{(1)}$	$x^{(2)}$
0	0	0	0	0
1	0	0	1	1
0	0	1	1	1
1	0	1	0	0
0	1	0	1	0
1	1	0	0	1
0	1	1	0	1
1	1	1	1	0

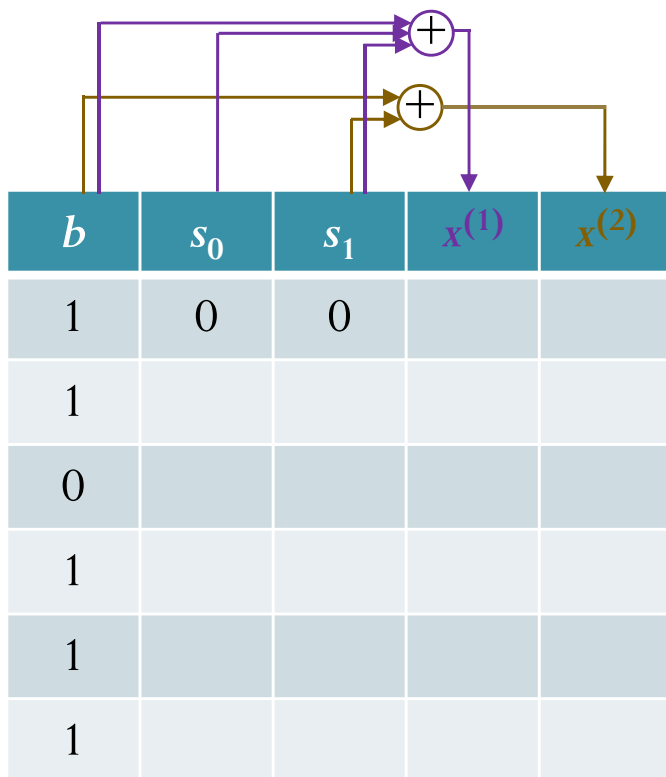


# Tracing the State Diagram to Find the Outputs

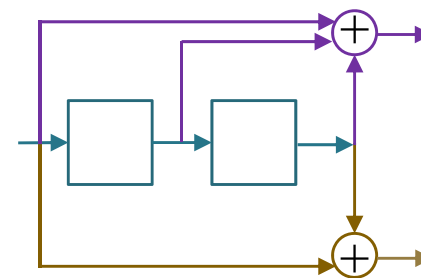
Input	1	1	0	1	1	1
Output	11	01	01	00	01	10



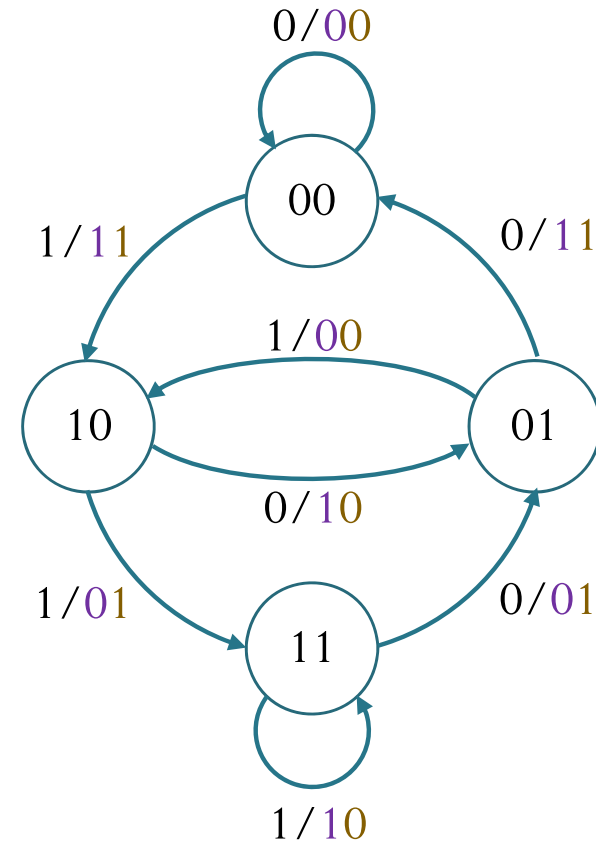
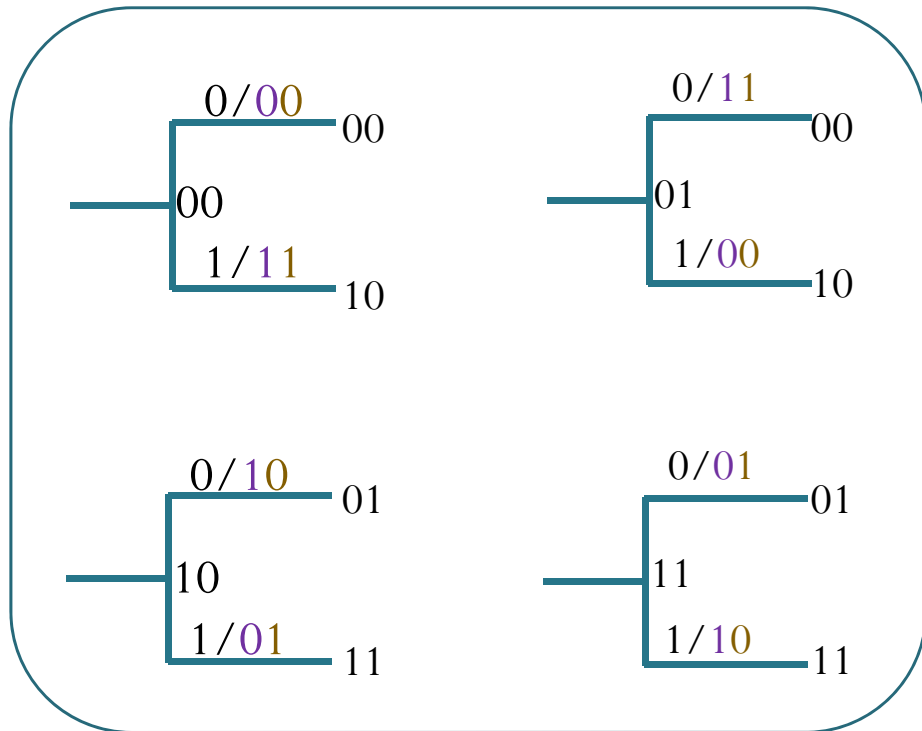
# Directly Finding the Output



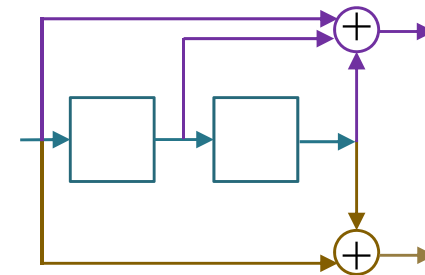
Input	1	1	0	1	1	1
Output						



# Parts for Code Tree



Two branches initiate from each node, the upper one for 0 and the lower one for 1.

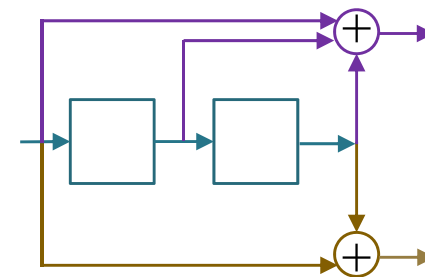
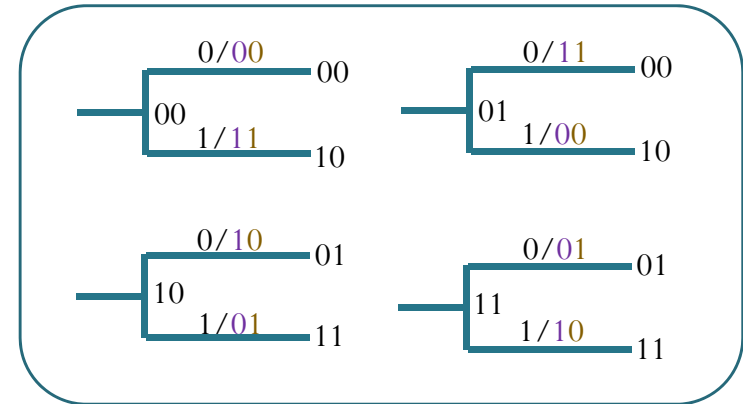
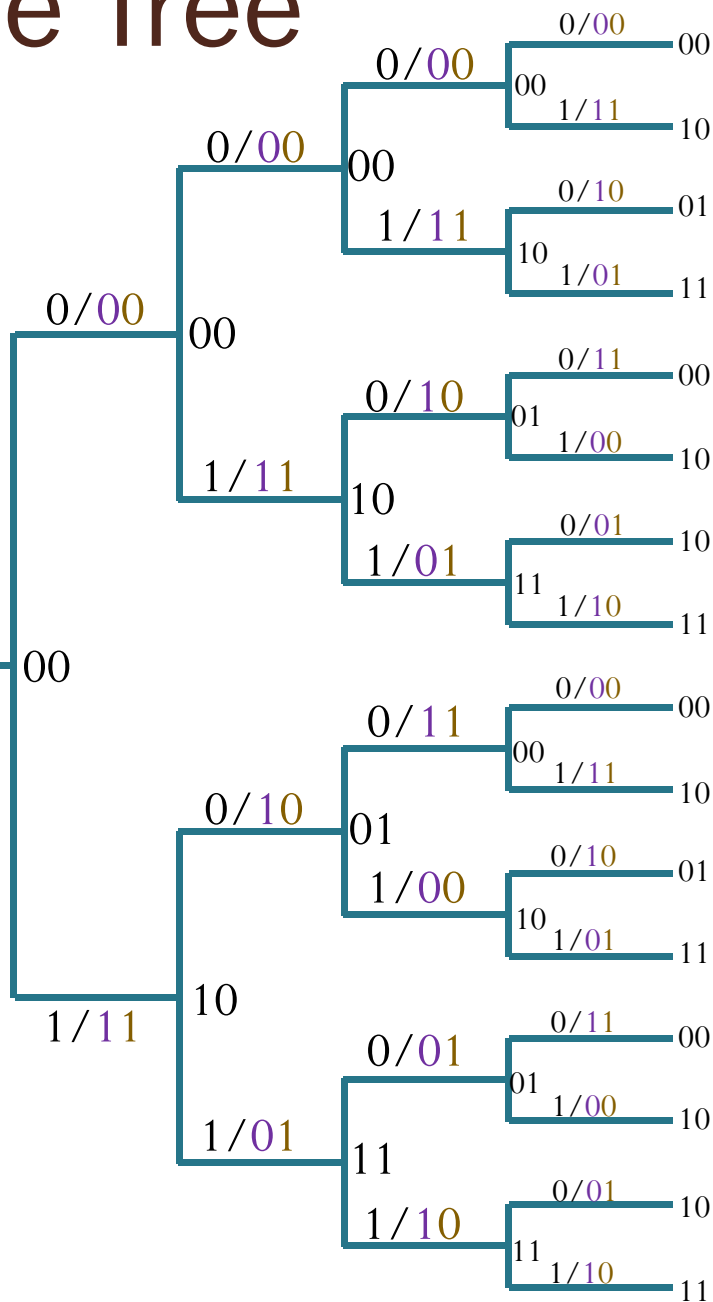


Show the coded output for any possible sequence of data digits.

# Code Tree

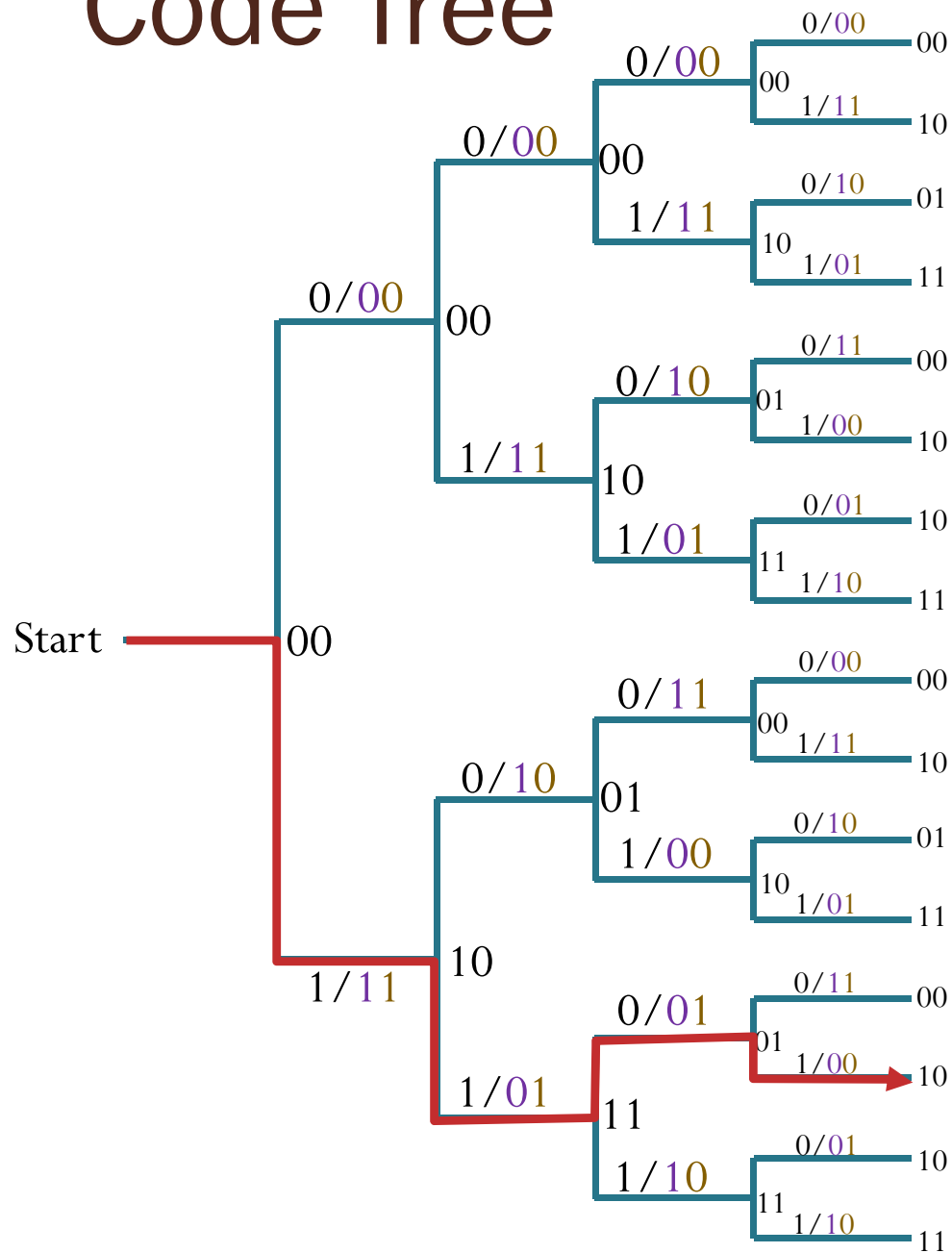
Initially, we always assume that all the contents of the register are 0.

Start

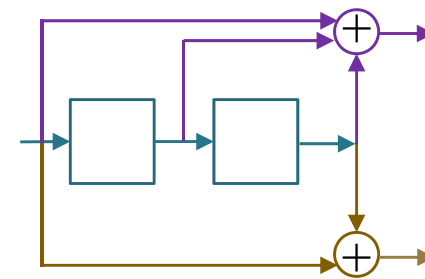




# Code Tree

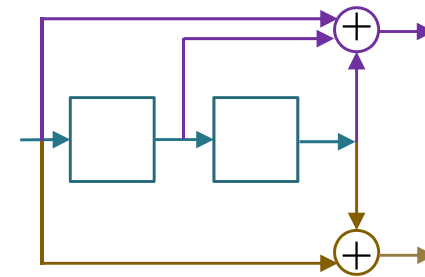
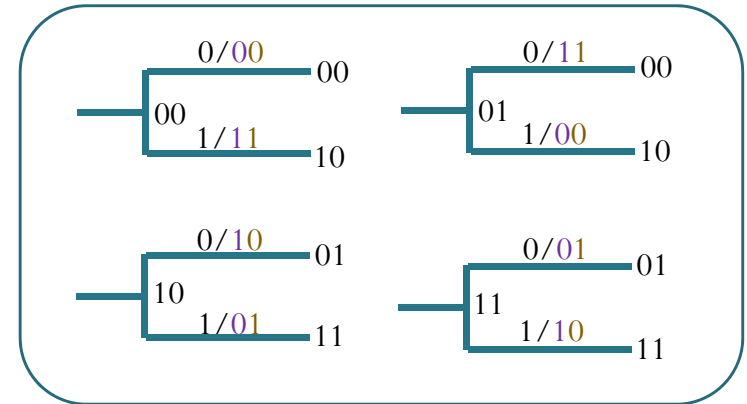
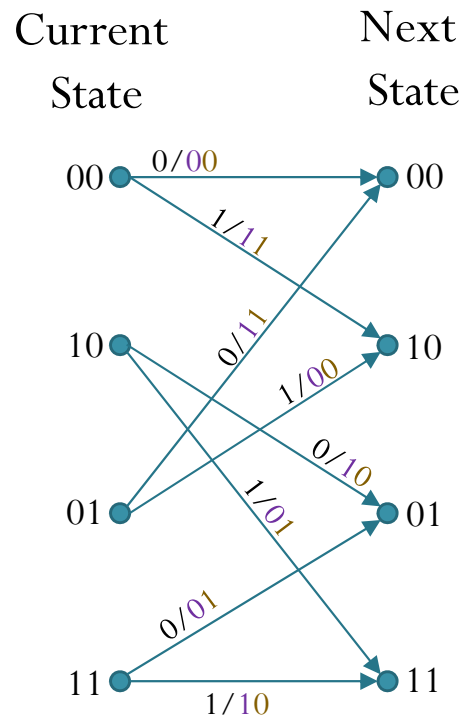


Input	1	1	0	1
Output	11	01	01	00



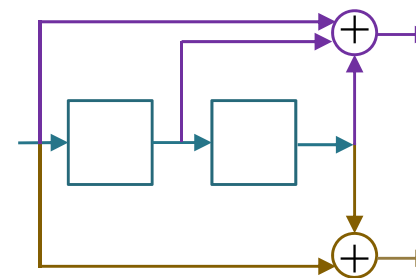
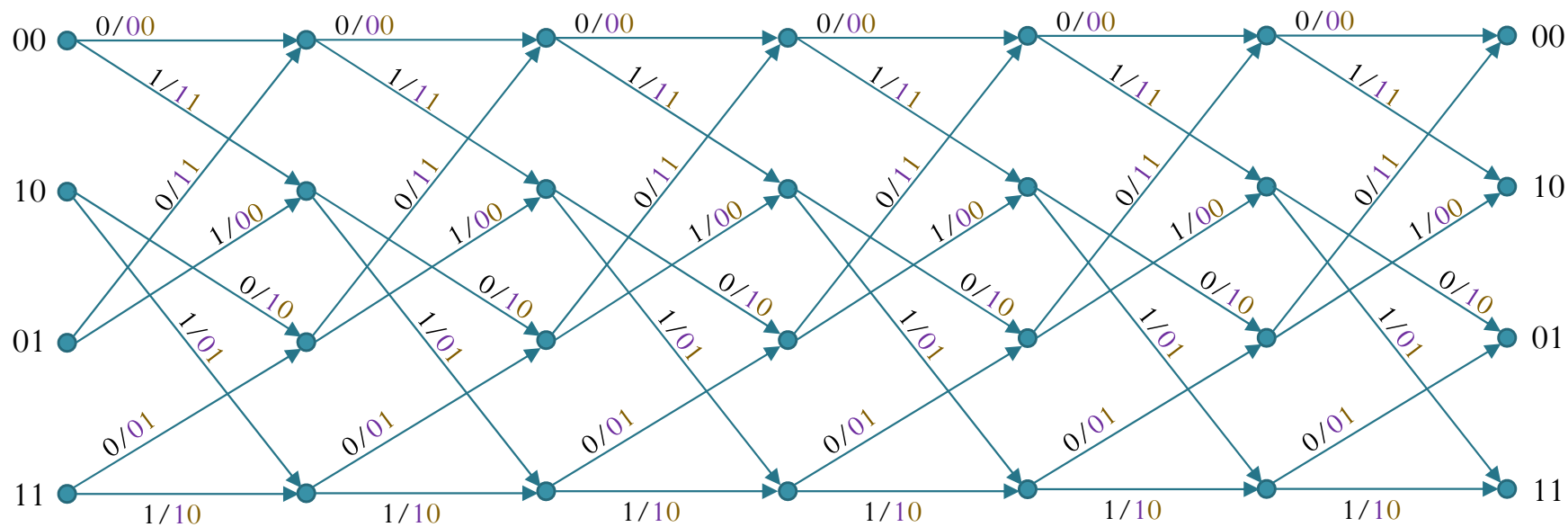
# Code Trellis

[Carlson & Crilly, 2009, p. 620]

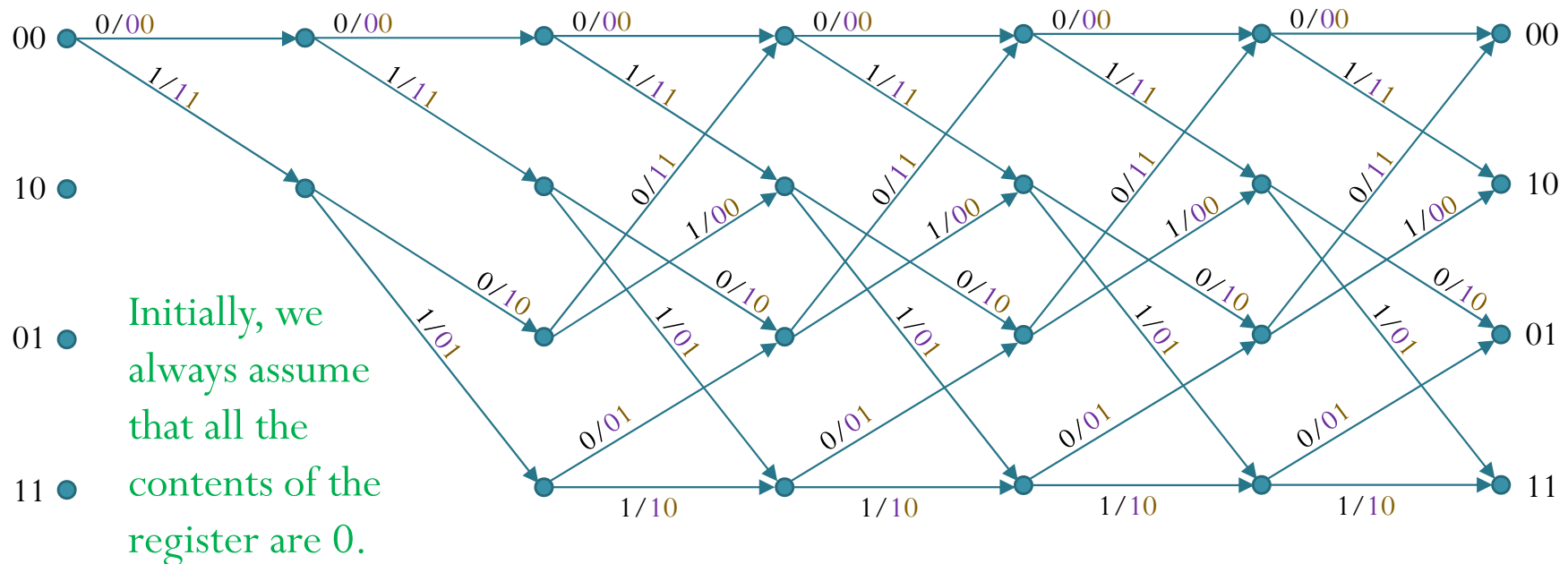


# Towards the Trellis Diagram

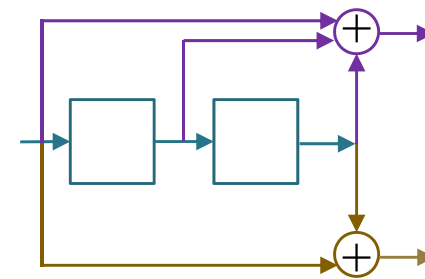
Another useful way of representing the code tree.



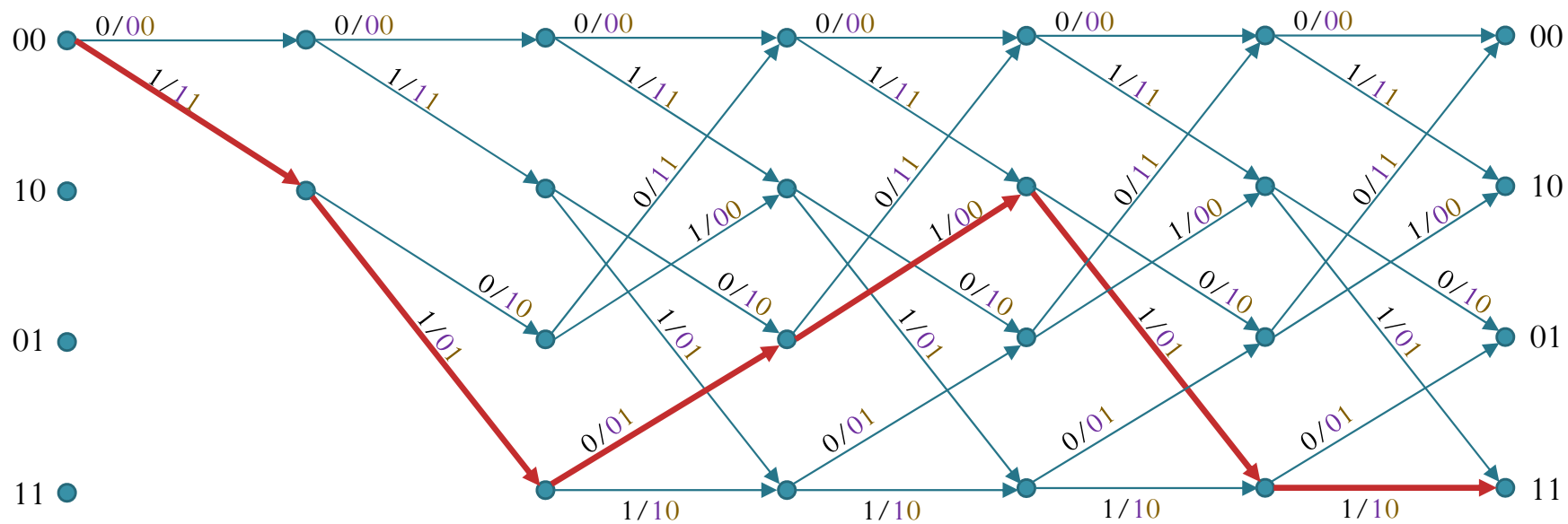
# Trellis Diagram



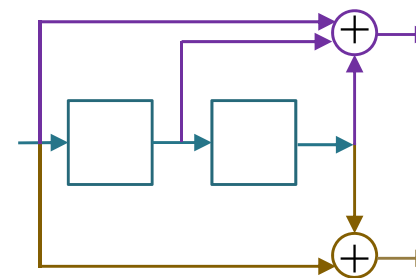
Each path that traverses through the trellis represents a valid codeword.



# Trellis Diagram

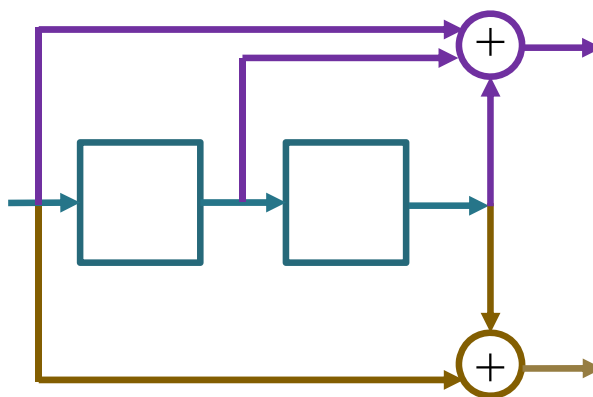


Input	1	1	0	1	1	1
Output	11	01	01	00	01	10



# Direct Minimum Distance Decoding

- Suppose  $\underline{\mathbf{y}} = [11\ 01\ 11]$ .
- Find  $\underline{\hat{\mathbf{b}}}$ .
  - Find the message  $\underline{\hat{\mathbf{b}}}$  which corresponds to the (valid) codeword  $\underline{\hat{\mathbf{x}}}$  with **minimum (Hamming) distance** from  $\underline{\mathbf{y}}$ .
    - $\underline{\hat{\mathbf{x}}} = \arg \min_{\underline{\mathbf{x}}} d(\underline{\mathbf{x}}, \underline{\mathbf{y}})$

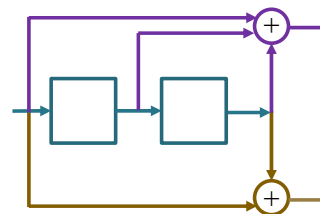


# Direct Minimum Distance Decoding

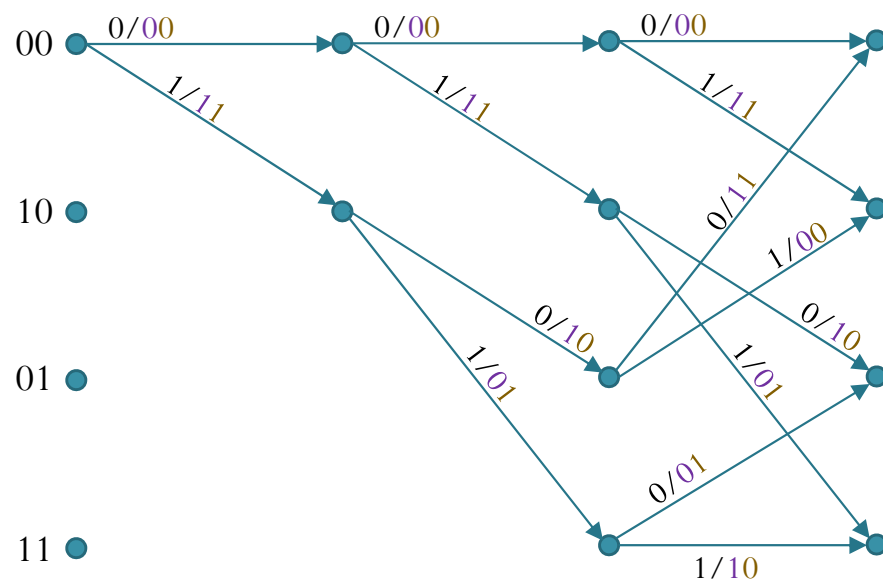
- Suppose  $\mathbf{y} = [11\ 01\ 11]$ .

- Find  $\hat{\mathbf{b}}$ .

- Find the message  $\hat{\mathbf{b}}$  which corresponds to the (valid) codeword  $\hat{\mathbf{x}}$  with minimum (Hamming) distance from  $\mathbf{y}$ .



$\mathbf{y} = [ \quad 11 \quad \quad 01 \quad \quad 11 \quad ]$ .

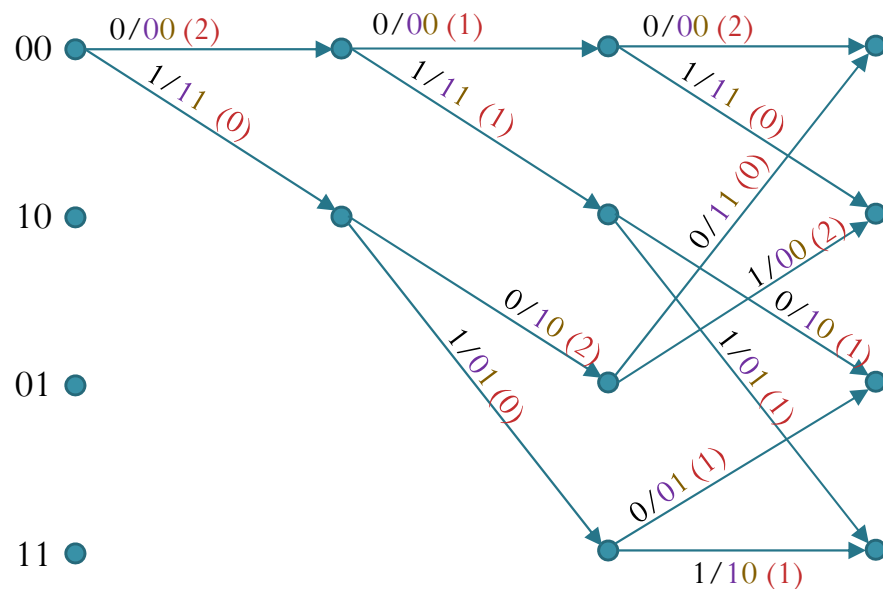


For 3-bit message,  
there are  $2^3 = 8$  possible codewords.  
We can list all possible codewords.  
However, here, let's first try to work  
on the distance directly.

# Direct Minimum Distance Decoding

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$\mathbf{y} = [ \quad 11 \quad \quad 01 \quad \quad 11 \quad ]$ .



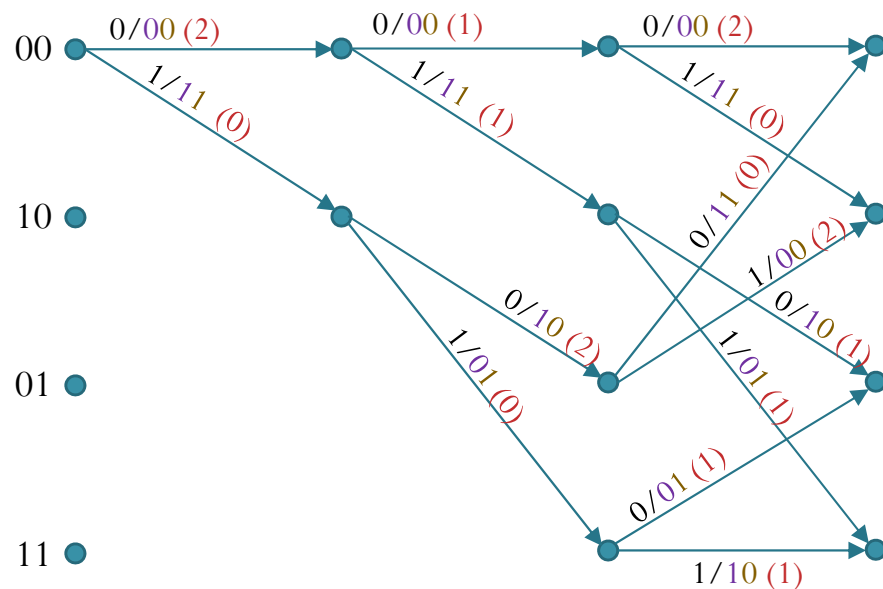
The number in parentheses on each branch is the branch metric, obtained by counting the differences between the encoded bits and the corresponding bits in  $\mathbf{y}$ .



# Direct Minimum Distance Decoding

- Suppose  $\mathbf{y} = [11\ 01\ 11]$ .
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  - Find the message  $\hat{\mathbf{b}}$  which corresponds to the (valid) codeword  $\hat{\mathbf{x}}$  with minimum (Hamming) distance from  $\mathbf{y}$ .

$\mathbf{y} = [ \quad 11 \quad \quad 01 \quad \quad 11 \quad ]$ .



$\mathbf{b}$	$d(\mathbf{x}, \mathbf{y})$
000	2+1+2 = 5
001	2+1+0 = 3
010	2+1+1 = 4
011	2+1+1 = 4
100	0+2+0 = 2
101	0+2+2 = 4
110	0+0+1 = 1
111	0+0+1 = 1

# Viterbi decoding

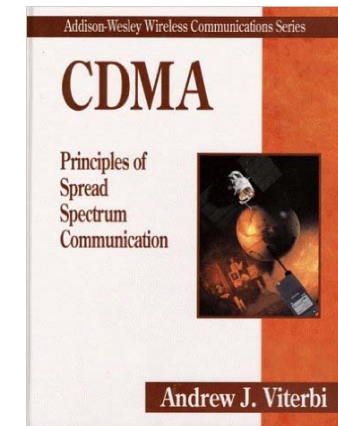
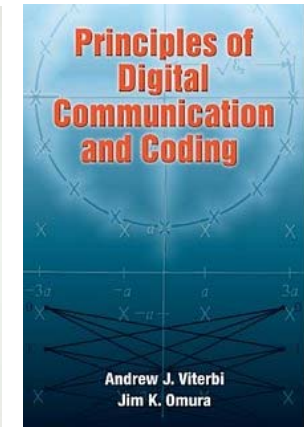
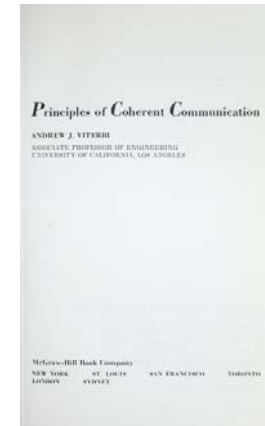
- Developed by Andrew J. Viterbi
  - Also co-founded Qualcomm Inc.
- Published in the paper “Error Bounds for Convolutional Codes and an Asymptotically Optimum Decoding Algorithm”, IEEE Transactions on Information Theory, Volume IT-13, pages 260-269, in April, **1967**.



[https://en.wikipedia.org/wiki/Andrew\\_Viterbi](https://en.wikipedia.org/wiki/Andrew_Viterbi)

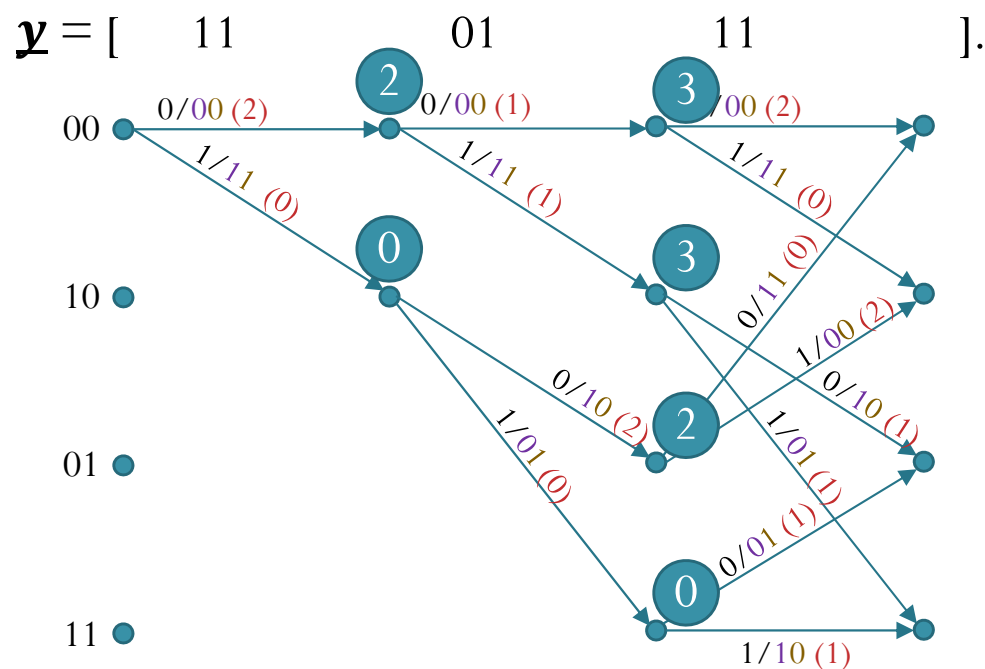
# Andrew J. Viterbi

- 1991: Claude E. Shannon Award
- 1952-1957: MIT BS & MS
  - Studied electronics and communications theory under such renowned scholars as Norbert Wiener, Claude Shannon, Bruno Rossi and Roberto Fano.
- 1962: Earned one of the first doctorates in electrical engineering granted at the University of Southern California (USC)
  - Ph.D. dissertation: error correcting codes
- 2004: USC Viterbi School of Engineering  
named in recognition of his \$52 million gift



# Viterbi Decoding: Ex. 1

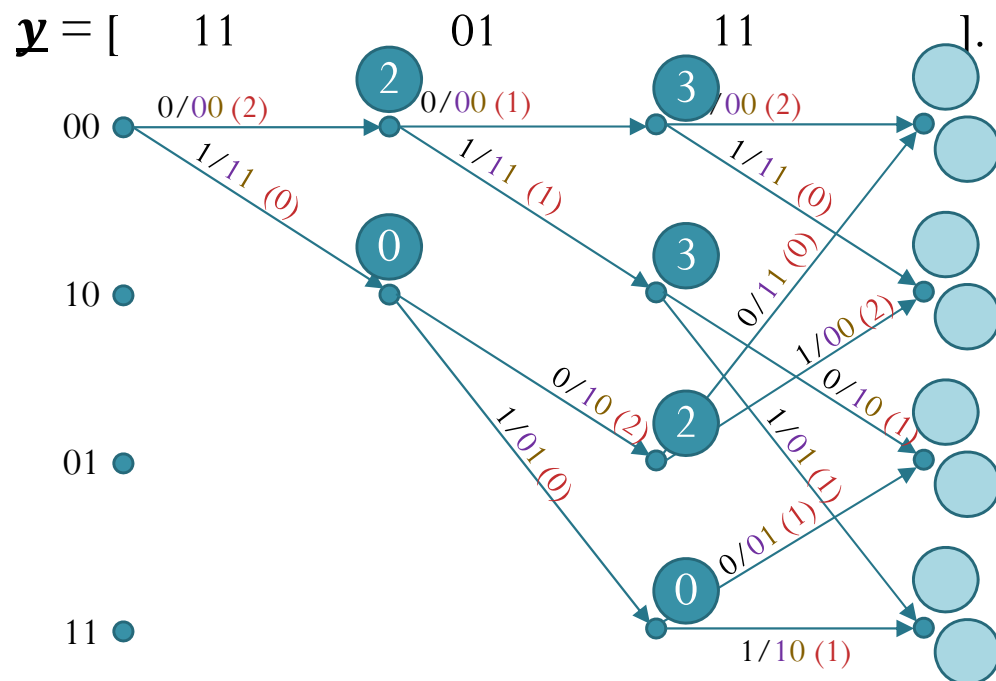
- Suppose  $\mathbf{y} = [11 \ 01 \ 11]$ .
- Find  $\hat{\mathbf{b}}$ .
  - Find the message  $\hat{\mathbf{b}}$  which corresponds to the (valid) codeword  $\hat{\mathbf{x}}$  with minimum (Hamming) distance from  $\mathbf{y}$ .



Each **circled number** at a node is the **running (cumulative) path metric**, obtained by summing branch metrics (distance) up to that node. Here, it is simply the **cumulative distance**.

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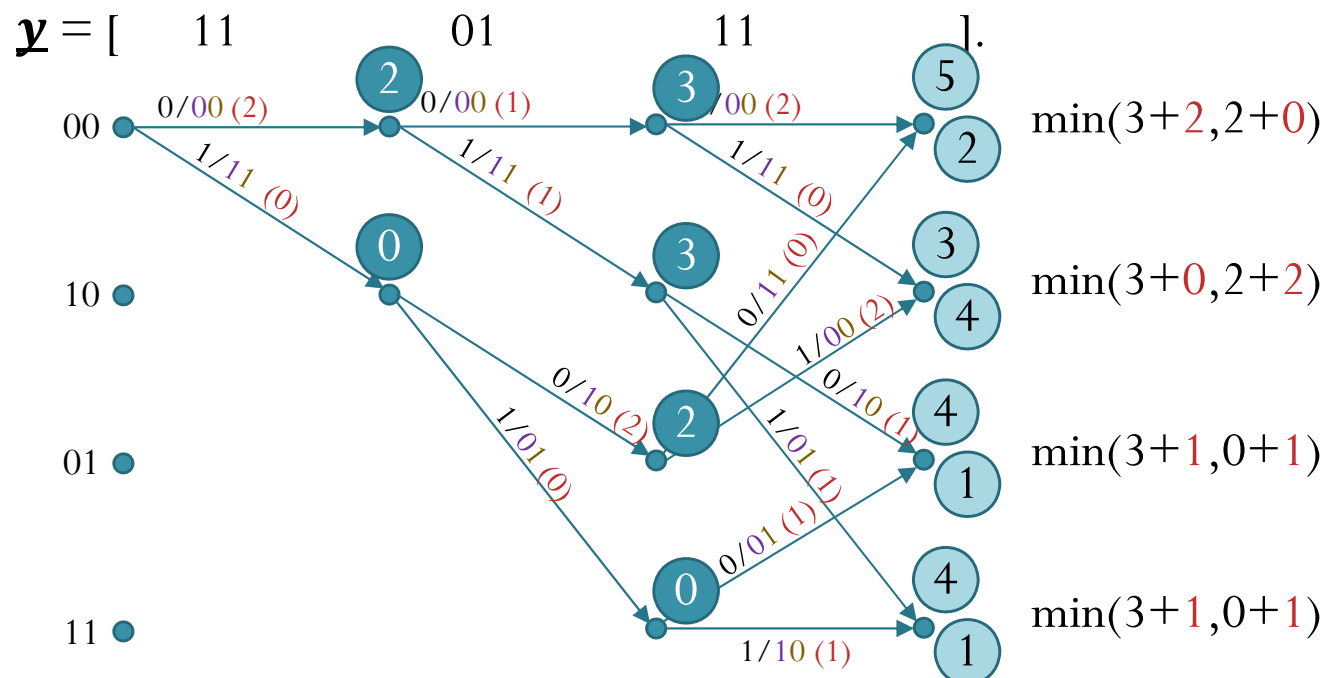
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- For the last column of nodes, each of the nodes has two branches going into it.
- So, there are two possible cumulative distance values.

# Viterbi Decoding: Ex. 1

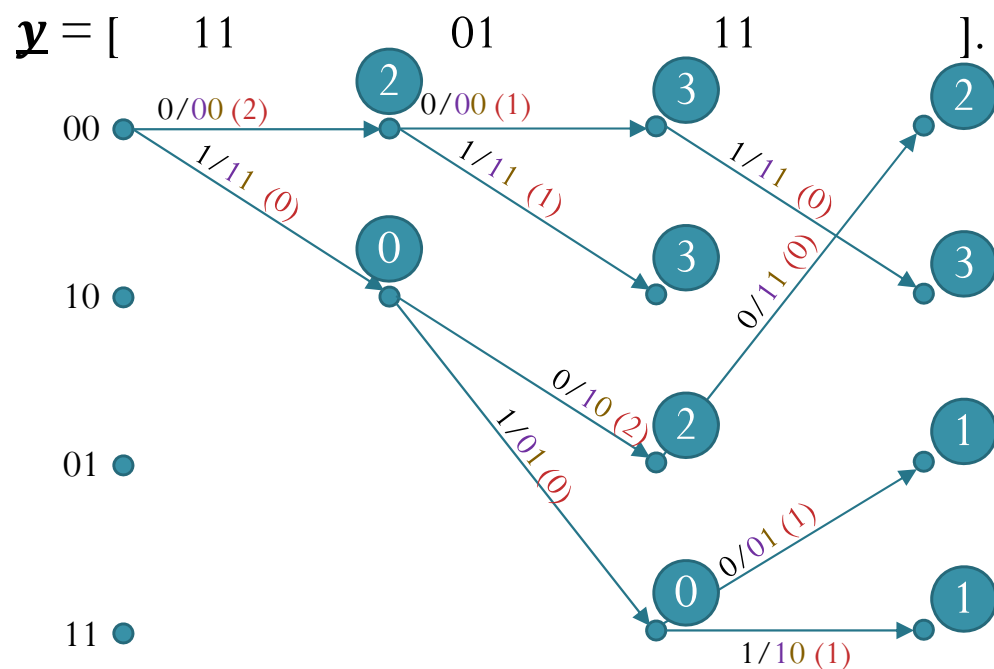
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We **discard the larger-distance path** because, regardless of what happens subsequently, this path will have a larger Hamming distance from  $\mathbf{y}$ .

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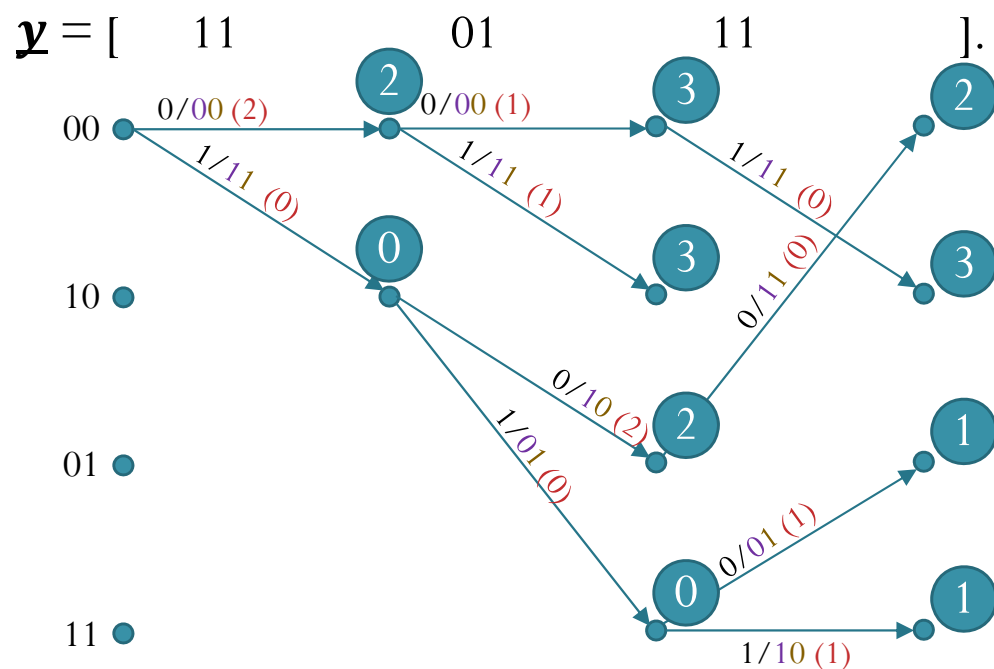
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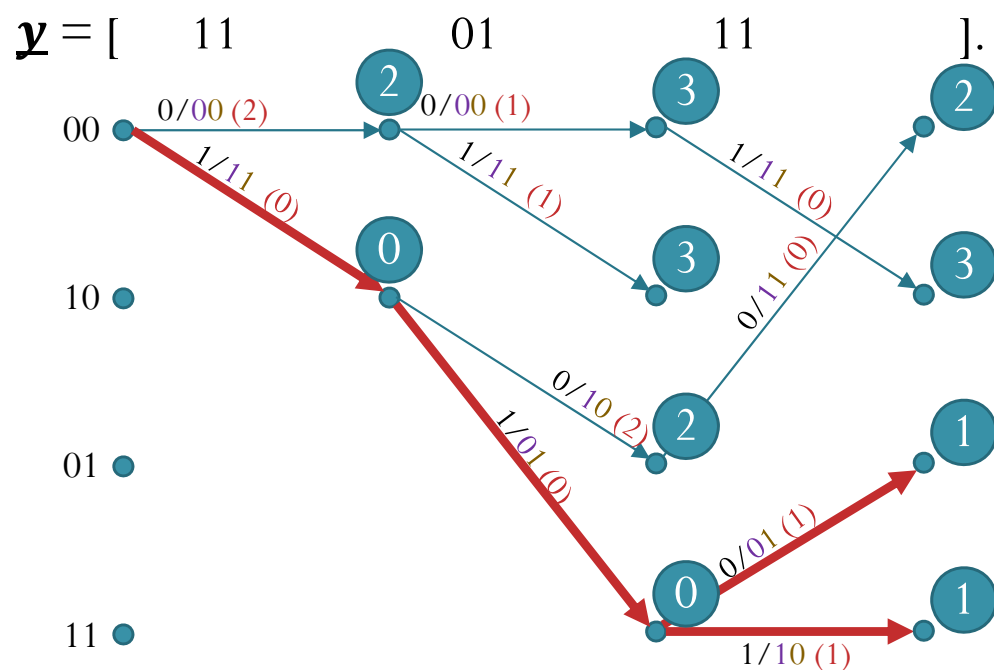


Note that we keep exactly one (optimal) **survivor path** to each state. (Unless there is a tie, then we keep both or choose any.)



# Viterbi Decoding: Ex. 1

- Suppose  $\mathbf{y} = [11\ 01\ 11]$ .
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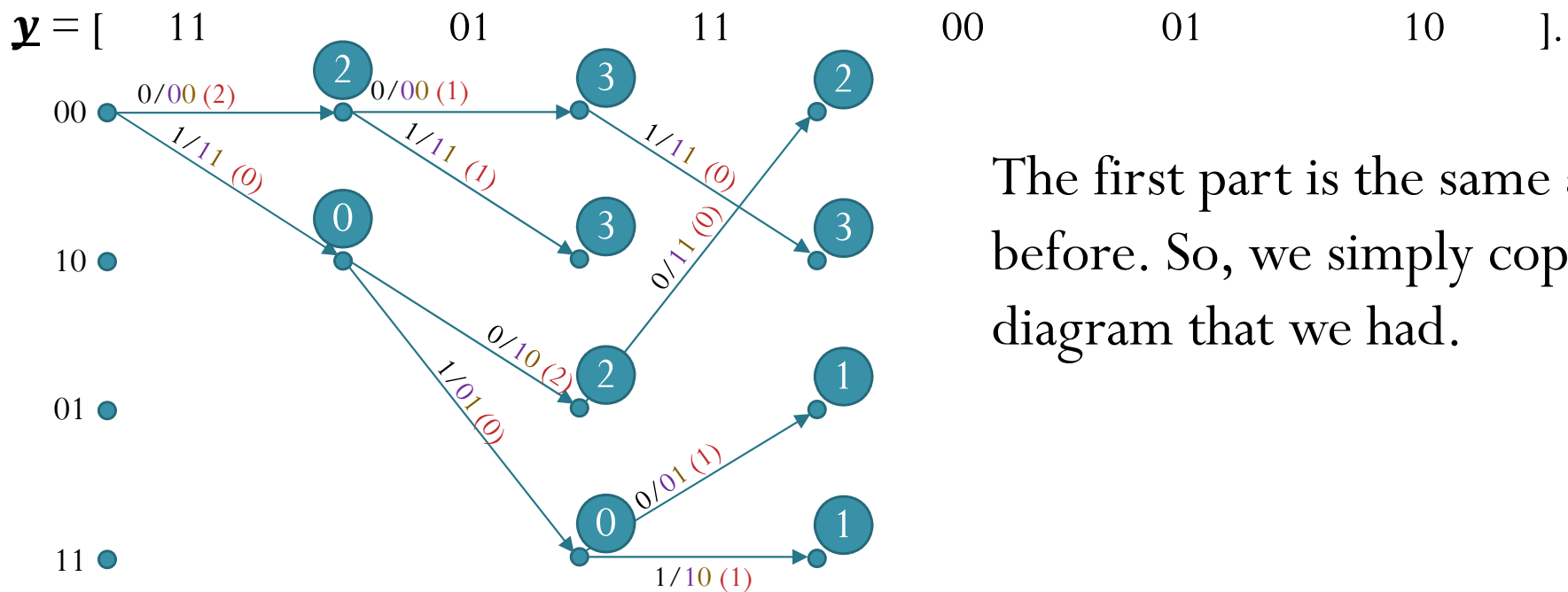


- So, the codewords which are nearest to  $\mathbf{y}$  is  $[11\ 01\ 01]$  or  $[11\ 01\ 10]$ .
- The corresponding messages are  $[110]$  or  $[111]$ , respectively.

# Viterbi Decoding: Ex. 2

- Suppose  $\mathbf{y} = [11 \ 01 \ 11 \ 00 \ 01 \ 10]$ .
- Find  $\hat{\mathbf{b}}$ .

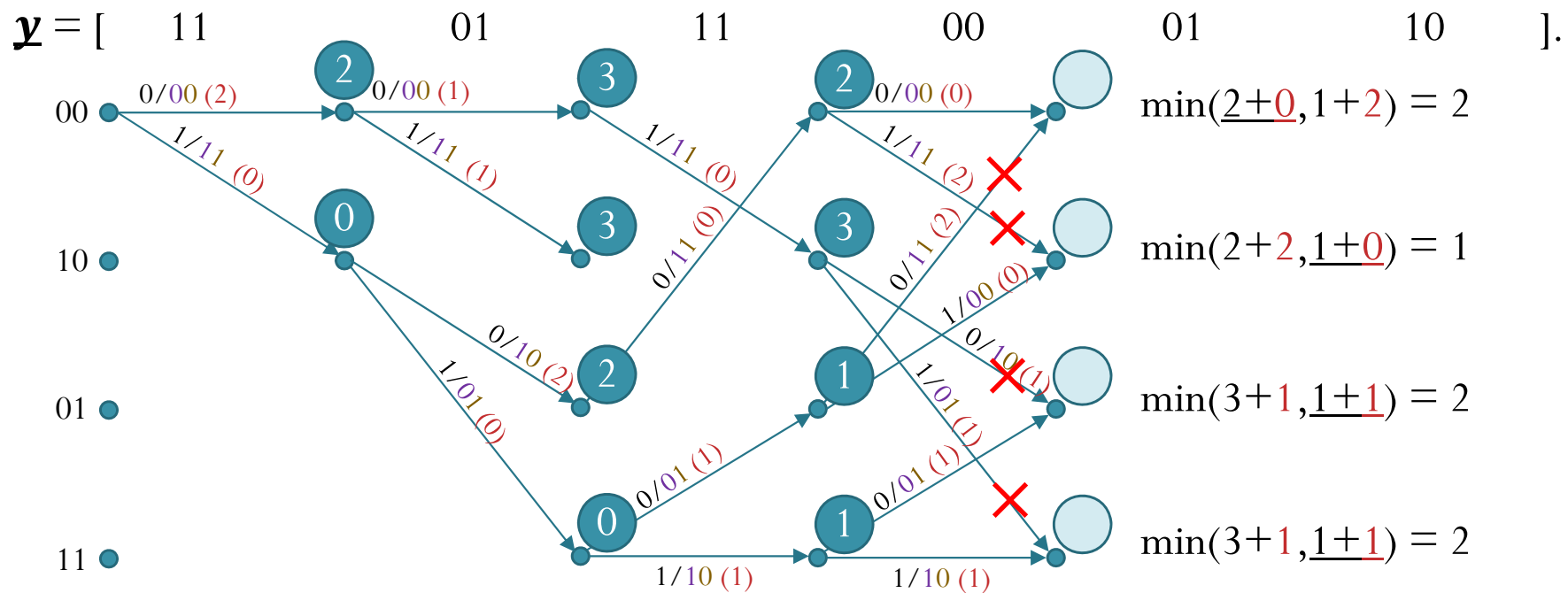
same as before



The first part is the same as before. So, we simply copy the diagram that we had.

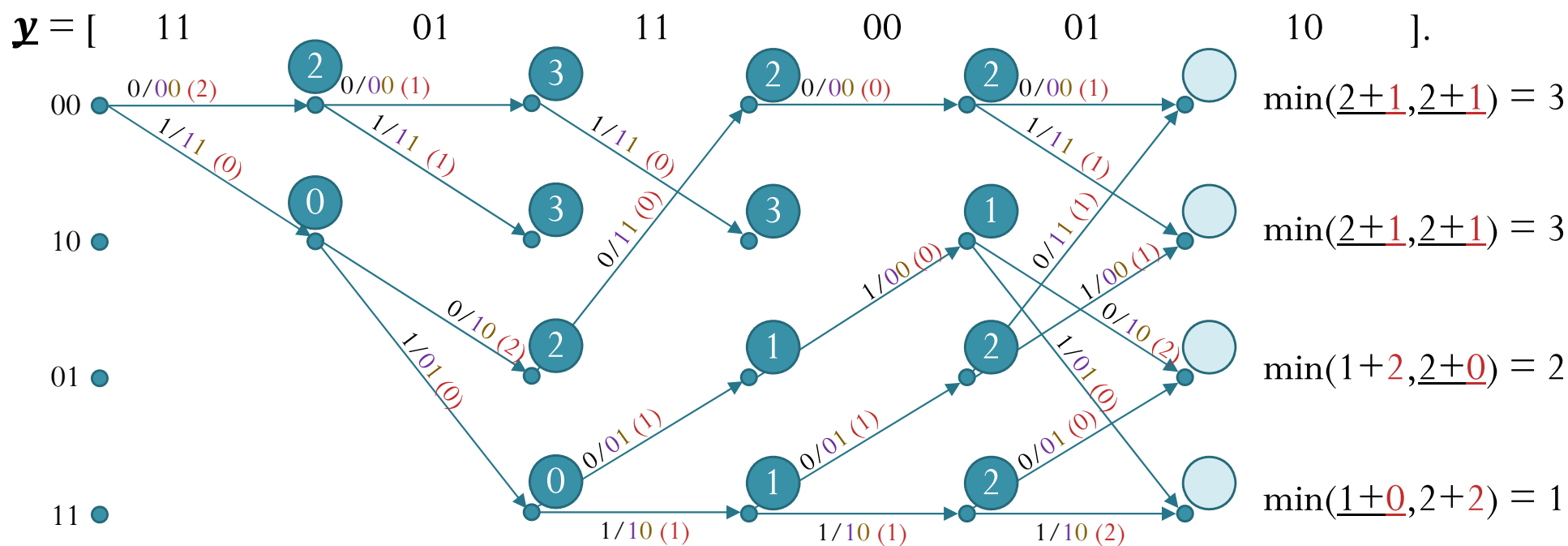
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- Suppose  $\mathbf{y} = [11 \ 01 \ 11 \ 00 \ 01 \ 10]$ .
- Find  $\hat{\mathbf{b}}$ .



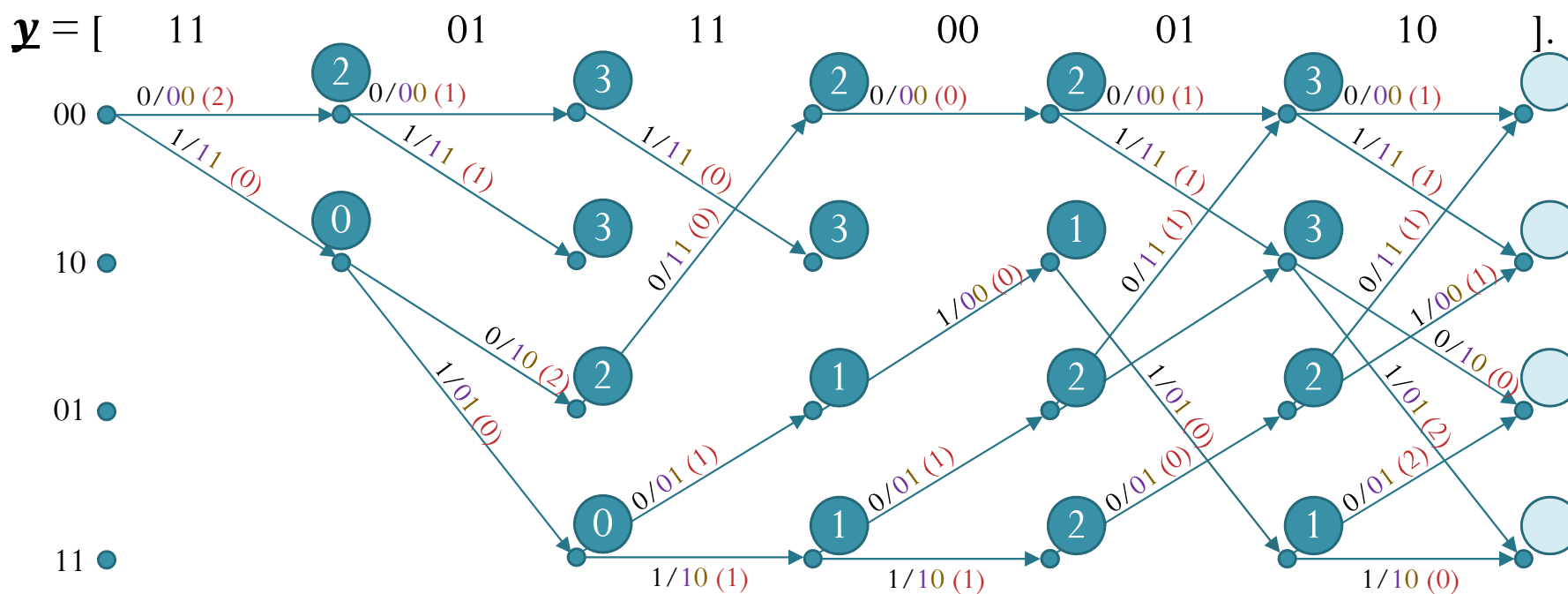
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- Suppose  $\mathbf{y} = [11 \ 01 \ 11 \ 00 \ 01 \ 10]$ .
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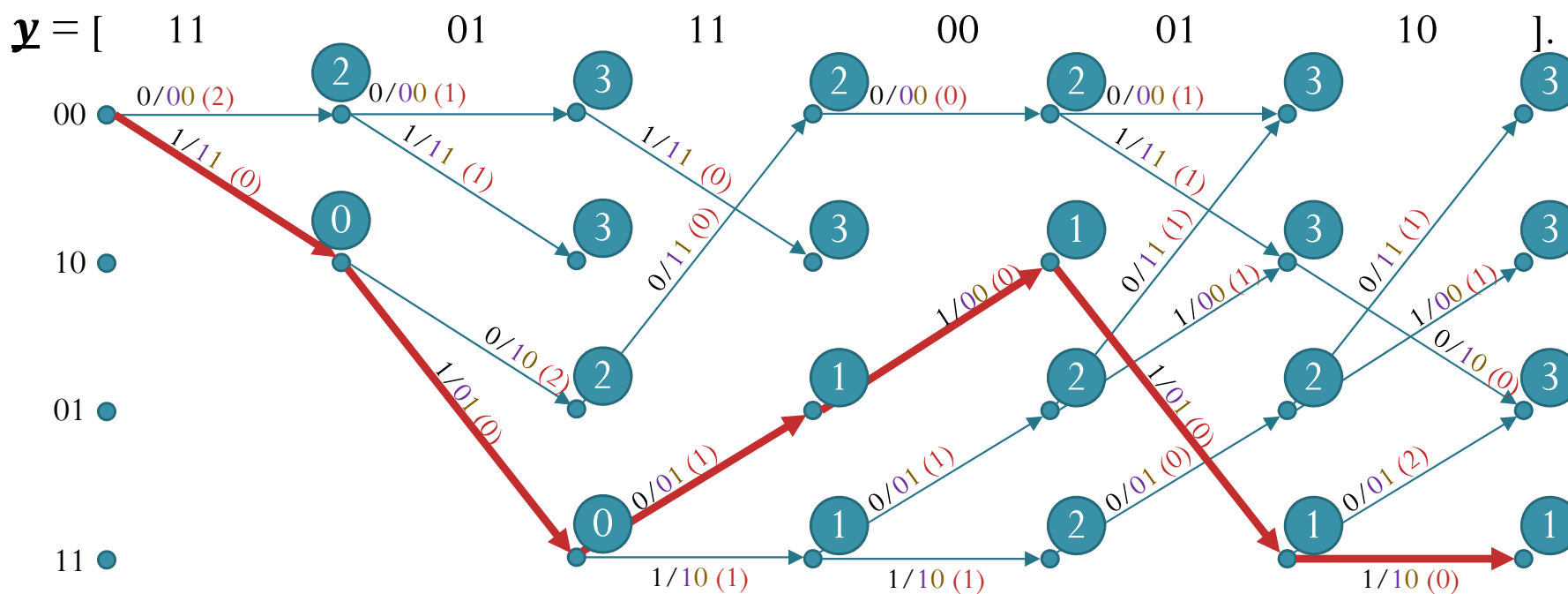
# Viterbi Decoding: Ex. 2

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- Find  $\hat{\mathbf{b}}$ .



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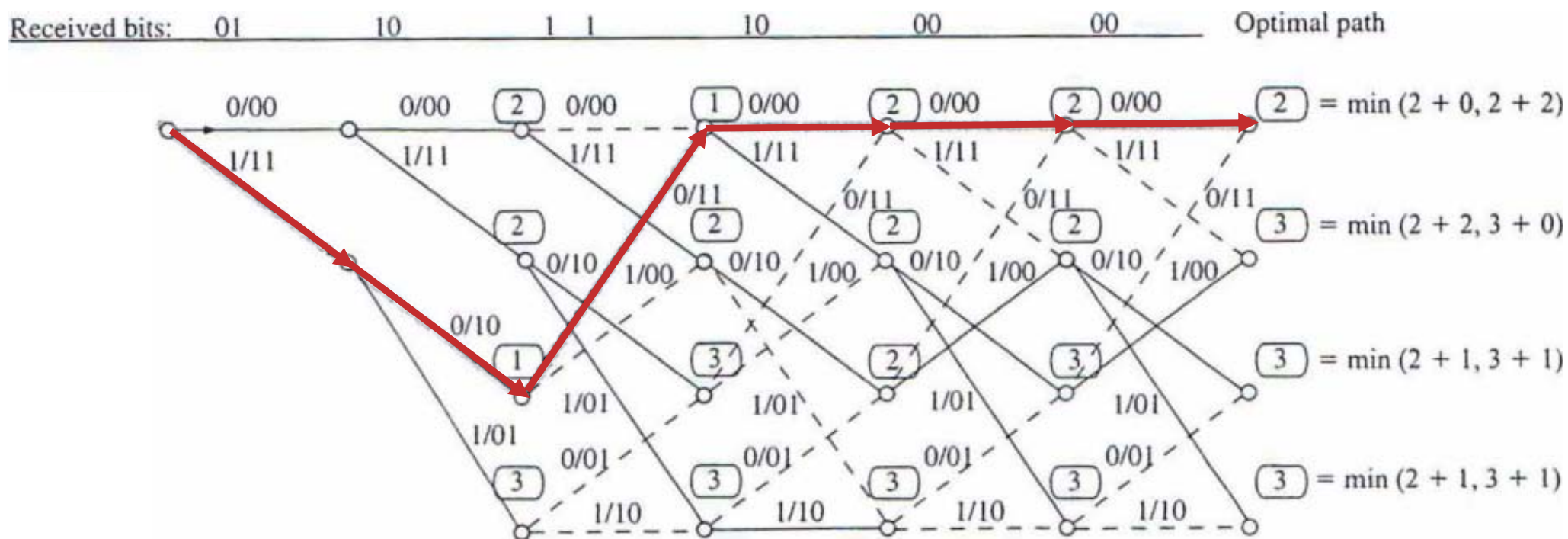


$$\hat{\mathbf{x}} = [11\ 01\ 01\ 00\ 01\ 10]$$

$$\hat{\mathbf{b}} = [1\ 1\ 0\ 1\ 1\ 1]$$

# Viterbi Decoding: Ex. 3

- Suppose  $\mathbf{y} = [01\ 10\ 11\ 10\ 00\ 00]$ .



$$\hat{\mathbf{x}} = [11\ 10\ 11\ 00\ 00\ 00]$$

$$\hat{\mathbf{b}} = [1\ 0\ 0\ 0\ 0\ 0]$$

# References: Conv. Codes

- Lathi and Ding, *Modern Digital and Analog Communication Systems*, 2009
  - [TK5101 L333 2009]
  - Section 15.6 p. 932-941
- Carlson and Crilly, *Communication Systems: An Introduction to Signals and Noise in Electrical Communication*, 2010
  - [TK5102.5 C3 2010]
  - Section 13.3 p. 617-637

